

## 100 mA Low-Dropout Voltage Regulator

#### **Features**

- · High Accuracy 5V, Guaranteed 100 mA Output
- · Extremely Low Quiescent Current
- · Low-Dropout Voltage
- · Extremely Tight Load and Line Regulation
- · Very Low Temperature Coefficient
- · Use as Regulator or Reference
- Needs Only 1 µF for Stability
- · Current and Thermal Limiting
- · Error Flag Warns of Output Dropout
- · Logic-Controlled Electronic Shutdown
- Output Programmable from 1.24V to 29V

#### **Applications**

- · Automotive Electronics
- · Voltage Reference
- Avionics

#### **General Description**

The LP2951 is micropower voltage regulators with very low dropout voltage (typically 40 mV at light loads and 380 mV at 100 mA), and very low quiescent current (75  $\mu$ A typical). The quiescent current of the LP2951 increases only slightly in dropout, thus prolonging battery life. This feature, among others, makes the LP2951 ideally suited for use in battery-powered systems.

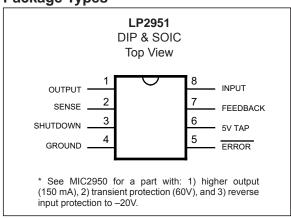
Available system functions, such as programmable output voltage and logic-controlled shutdown, are available as well.

Additional features available with the LP2951 also include an error flag output that warns of a low output voltage, which is often due to failing batteries on the input. This may also be used as a power-on reset. A logic-compatible shutdown input is also available which enables the regulator to be switched on and off. This part may also be pin-strapped for a 5V output, or programmed from 1.24V to 29V with the use of two external resistors.

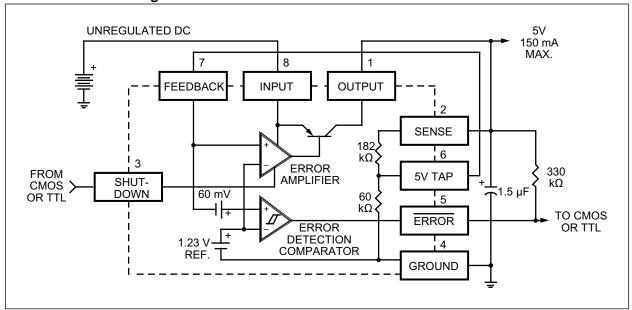
The LP2951 is available as either a -02 or -03 version. The -02 and -03 versions are guaranteed for junction temperatures from -40°C to +125°C; the -02 version has a tighter output and reference voltage specification range over temperature.

The LP2951 have a tight initial tolerance (0.5% typical), a very low output voltage temperature coefficient which allows use as a low-power voltage reference, and extremely good load and line regulation (0.05% typical). This greatly reduces the error in the overall circuit, and is the result of careful design techniques and process control.

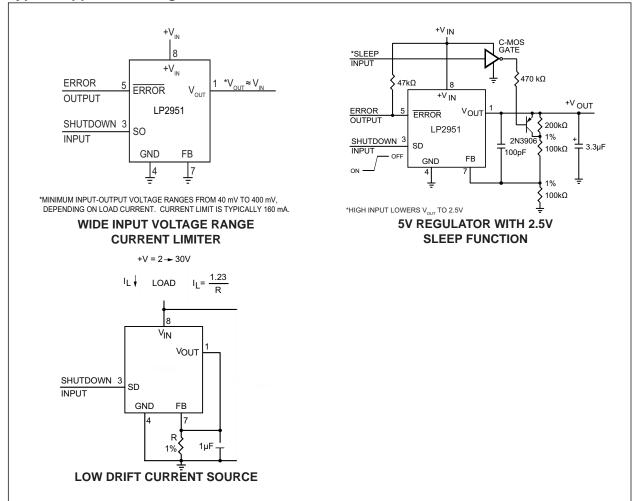
#### **Package Types**



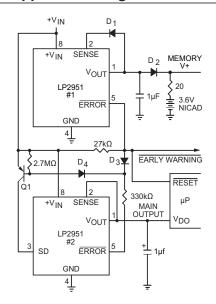
#### **Functional Block Diagram**



#### **Typical Application Diagrams**



#### **Typical Application Diagrams**

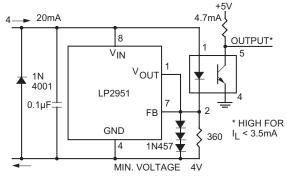


- $^{+}V_{IN}$ 8 \$470k +V<sub>IN</sub> Vout Vout **ERROR** 470k LP2951 FΒ 1μF SD RESET  $R_2$ **GND** 
  - LATCH-OFF WHEN ERROR FLAG OCCURS

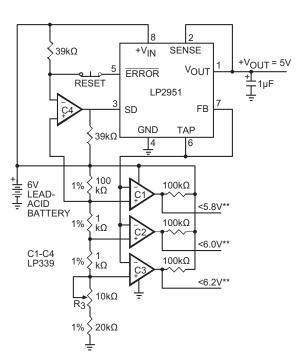
- EARLY WARNING FLAG ON LOW INPUT VOLTAGE
   MAIN OUTPUT LATCHES OFF AT LOWER INPUT VOLTAGES
   BATTERY BACKUP ON AUXILIARY OUTPUT

OPERATION: REG. #1'S V  $_{\rm OUT}$  IS PROGRAMMED ONE DIODE DROP ABOVE 5V. ITS ERROR FLAG BECOMES ACTIVE WHEN V  $_{\rm N}$   $\le$  5.7V. WHEN V  $_{\rm N}$  DROPS BELOW 5.3V, THE ERROR FLAG OF REG. #2 BECOMES ACTIVE AND VIA Q1 LATCHES THE MAIN OUTPUT OFF. WHEN  $\rm V_{in}$  AGAIN EXCEEDS 5.7V REG. #1 IS BACK IN REGULATION AND THE EARLY WARNING SIGNAL RISES,

#### **REGULATOR WITH EARLY WARNING** AND AUXILIARY OUTPUT

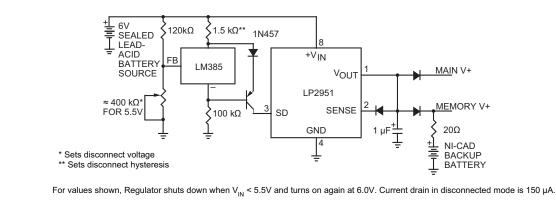


#### **OPEN CIRCUIT DETECTOR FOR 4 mA** TO 20 mA CURRENT LOOP

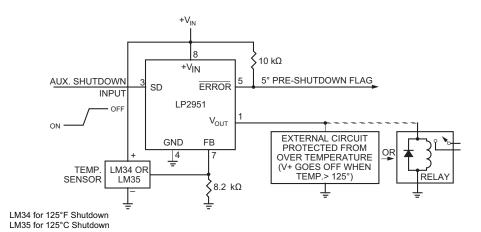


- C1 TO C4 ARE COMPARATORS (LP339 OR EQUIVALENT)
- \*OPTIONAL LATCH OFF WHEN DROP OUT OCCURS. ADJUST R3 FOR C2 SWITCHING WHEN  $\mathrm{V}_{\mathrm{IN}}$  IS  $6.0\mathrm{V}$
- \*\*OUTPUTS GO LOW WHEN  $V_{IN}$  DROPS BELOW DESIGNATED THRESHOLDS.

#### **REGULATOR WITH** STATE-OF-CHARGE INDICATOR



#### **LOW BATTERY DISCONNECT**



SYSTEM OVER TEMPERATURE **PROTECTION CIRCUIT** 

#### 1.0 ELECTRICAL CHARACTERISTICS

#### **Absolute Maximum Ratings † ††**

Power Dissipation	Internally Limited
Lead Temperature (soldering, 5 sec.)	
Storage Temperature	
Operating Junction Temperature Range(Note 1)	
LP2951	40°C to +125°C
Input Supply Voltage	
Feedback Input Voltage(Note 2, 3)	
Shutdown Input Voltage(Note 2)	
Error Comparator Output Voltage(Note 2)	

**† Notice:** Boldface limits apply at temperature extremes.

†† Notice: If Military/Aerospace specified devices are required, contact your local representative/distributor for availability and specifications.

- **Note 1:** The thermal resistance of the 8-pin DIP package is 105°C/W junction-to-ambient when soldered directly to a PC board. Junction-to-ambient thermal resistance for the SOIC (M) package is 160°C/W.
  - 2: May exceed input supply voltage.
  - **3:** When used in dual-supply systems where the output terminal sees loads returned to a negative supply, the output voltage should be diode-clamped to ground.

#### **ELECTRICAL CHARACTERISTICS**

Electrical Characteristics: Unless otherwise indicated, T <sub>A</sub> = +25°C								
Parameters Sym. Min. Typ. Max. Units Conditions								
Output Voltage T <sub>J</sub> = 25°C		4.975	5.000	5.025	V	LP2951-02 (±0.5%)		
		4.950	5.000	5.050	V	LP2951-03 (±1%)		
		4.802	4.850	4.899	V	LP2951-4.8 (±1%)		

- **Note 1:** Output or reference voltage temperature coefficient is defined as the worst case voltage change divided by the total temperature range.
  - **2:** Regulation is measured at constant junction temperature, using pulse testing with a low duty cycle. Changes in output voltage due to heating effects are covered in the specification for thermal regulation.
  - 3: Line regulation for the LP2951 is tested at 150°C for  $I_L$  = 1 mA. For  $I_L$  = 100  $\mu$ A and  $T_J$  = 125°C, line regulation is guaranteed by design to 0.2%. See Typical Performance Characteristics for line regulation versus temperature and load current.
  - **4:** Dropout voltage is defined as the input to output differential at which the output voltage drops 100 mV below its nominal value measured at 1V differential. At very low values of programmed output voltage, the minimum input supply voltage of 2V (2.3V over temperature) must be taken into account.
  - 5: Thermal regulation is defined as the change in output voltage at a time T after a change in power dissipation is applied, excluding load or line regulation effects. Specifications are for a 50 mA load pulse at V<sub>IN</sub> = 30V (1.25W pulse) for t = 10 ms.
  - **6:** Comparator thresholds are expressed in terms of a voltage differential at the Feedback terminal below the nominal reference voltage measured at 6V input. To express these thresholds in terms of output voltage change, multiply by the error amplifier gain = V<sub>OUT</sub>/V<sub>REF</sub> =(R1 + R2)/R2. For example, at a programmed output voltage of 5V, the Error output is guaranteed to go low when the output drops by 95 mV x 5V/1.235V = 384 mV. Thresholds remain constant as a percent of V<sub>OUT</sub> as V<sub>OUT</sub> is varied, with the dropout warning occurring at typically 5% below nominal, 7.5% guaranteed.
  - 7:  $V_{REF} \le V_{OUT} \le (V_{IN} 1 V)$ ,  $2.3V \le V_{IN} \le 30V$ ,  $100 \mu A < I_L \le 100 mA$ ,  $T_J \le T_{JMAX}$ .
  - **8:** Output or reference voltage temperature coefficient is defined as the worst case voltage change divided by the total temperature range.
  - 9: V<sub>SHUTDOWN</sub> ≥ 2V, V<sub>IN</sub> ≤ 30V, V<sub>OUT</sub> = 0, with Feedback pin tied to 5V Tap.

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Parameters	Sym.	Min.	Тур.	Max.	Units	Conditions			
		4.950	_	5.050	V	LP2951-02 (±0.5%)			
Output Voltage –25°C ≤ T <sub>.1</sub> ≤ +85°C		4.925	_	5.075	V	LP2951-03 (±1%)			
20 0 2 1 1 2 1 00 0		4.777	_	4.872	V	LP2951-4.8 (±1%)			
Output Voltage		4.940	_	5.060	V	LP2951-02 (±0.5%)			
Over Full Temperature Range		4.900	_	5.100	V	LP2951-03 (±1%)			
-40°C to +125°C		4.753	_	4.947	V	LP2951-4.8 (±1%)			
		4.930	_	5.070	V	LP2951-02 (±0.5%), 100 $\mu$ A $\leq I_L \leq 100$ mA, $T_J \leq T_{J(max)}$			
Output Voltage Over Load Variation		4.880	_	5.120	V	LP2951-03 (±1%), 100 $\mu$ A $\leq$ I <sub>L</sub> $\leq$ 100 mA, T <sub>J</sub> $\leq$ T <sub>J(max)</sub>			
		4.733	_	4.967	V	LP2951-4.8 (±1%), 100 µA ≤ I <sub>L</sub> ≤ 100 mA, T <sub>J</sub> ≤ T <sub>J(max)</sub>			
		_	20	100	ppm/ °C	LP2951-02 (±0.5%), Note 1			
Output Voltage Temperature Coefficient		_	50	150	ppm/ °C	LP2951-03 (±1%), Note 1			
		_	50	150	ppm/ °C	LP2951-4.8 (±1%), Note 1			

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  - 7:  $V_{REF} \le V_{OUT} \le (V_{IN} 1 \text{ V})$ , 2.3V  $\le V_{IN} \le 30 \text{ V}$ , 100  $\mu\text{A} < I_L \le 100 \text{ mA}$ ,  $T_J \le T_{JMAX}$ .
  - **8:** Output or reference voltage temperature coefficient is defined as the worst case voltage change divided by the total temperature range.
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Electrical Characteristics: Unless otherwise indicated, T <sub>A</sub> = +25°C									
Parameters	Sym.	Min.	Тур.	Max.	Units	Conditions			
		_	0.03	0.10	%	LP2951-02 (±0.5%), Note 2,			
		_	_	0.20	%	3			
Line Regulation		_	0.04	0.20	%	LP2951-03 (±1%), Note 2, 3			
Line Regulation		_	_	0.40	%	LP2931-03 (±1%), Note 2, 3			
		_	0.04	0.20	%	LP2951-4.8 (±1%), Note 2, 3			
		_	_	0.40	%	LP2951-4.6 (±1%), Note 2, 3			
		_	0.04	0.10	%	LP2951-02 (±0.5%), Note 2,			
		_	_	0.20	%	100 μA ≤ I <sub>L</sub> ≤ 100 mA			
Load Regulation		_	0.10	0.20	%	LP2951-03 (±1%), Note 2,			
Load Regulation		_	_	0.30	%	100 μA ≤ I <sub>L</sub> ≤ 100 mA			
		_	0.10	0.20	%	LP2951-4.8 (±1%), Note 2,			
		_	_	0.30	%	100 μA ≤ I <sub>L</sub> ≤ 100 mA			
		_	50	80	mV	Note 4, I <sub>1</sub> = 100 μA			
Dranaut Valtaga		_	_	150	mV	Νοίε 4, 1[ – 100 μΑ			
Dropout Voltage		_	380	450	mV	Note 4, I <sub>1</sub> = 100 mA			
		_	_	600	mV	Note 4, IL = 100 IIIA			
		_	100	150	μA	I <sub>I</sub> = 100 μA			
Ground Current		_	_	200	μA	- 100 μΑ			
Giouria Current		_	8	12	mA	- 100 mA			
		_	_	14	mA	I <sub>L</sub> = 100 mA			

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  - 7:  $V_{REF} \le V_{OUT} \le (V_{IN} 1 \ V)$ , 2.3 $V \le V_{IN} \le 30V$ , 100  $\mu$ A <  $I_L \le 100 \ m$ A,  $T_J \le T_{JMAX}$ .
  - 8: Output or reference voltage temperature coefficient is defined as the worst case voltage change divided by the total temperature range.
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Electrical Characteristics: Unless otherwise indicated, T <sub>A</sub> = +25°C										
Parameters	Sym.	Min.	Тур.	Max.	Units	Conditions				
Dropout Current		_	180	250 310	μA μA	V <sub>IN</sub> = 4.5V, I <sub>L</sub> = 100 μA				
Current Limit		_	160	220 220	mA mA	V <sub>OUT</sub> = 0V				
Thermal Regulation		_	0.05	0.20	%/W	Note 5				
		_	430	_	μV <sub>RMS</sub>	10 Hz to 100 kHz, $C_L = 1 \mu F$				
		_	160	_	μV <sub>RMS</sub>	10 Hz to 100 kHz, C <sub>L</sub> = 200 µF				
Output Noise		_	100	_	μV <sub>RMS</sub>	10 Hz to 100 kHz, C <sub>L</sub> = 3.3 μF, 0.01 μF bypass Feedback to Output				
		1.220	1.235	1.25	V	L D2054 02 (10 50/)				
		1.200	_	1.260	V	LP2951-02 (±0.5%)				
Deference Voltage		1.210	1.235	1.260	V	LP2951-03 (±1%)				
Reference Voltage		1.200	_	1.270	V	LP2951-03 (±1%)				
		1.210	1.235	1.260	V	L D2054 4.9 (140/)				
		1.200	_	1.270	V	LP2951-4.8 (±1%)				
		1.190	_	1.270	V	LP2951-02 (±0.5%), Note 7				
Reference Voltage		1.185	_	1.285	V	LP2951-03 (±1%), Note 7				
		1.185	_	1.285	V	LP2951-4.8 (±1%), Note 7				

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Parameters	Sym.	Min.	Тур.	Max.	Units	Conditions				
Facility of Diag Occurrent		_	20	40	A					
Feedback Bias Current		_	_	60	nA	_				
		_	20	_	ppm/°C	LP2951-02 (±0.5%), Note 8				
Reference Voltage		_	50	_	ppm/°C	LP2951-03 (±1%), Note 8				
		_	50	_	ppm/°C	LP2951-4.8 (±1%), Note 8				
Feedback Bias Current Temperature Coefficient		_	0.1	_	nA/°C					
Output Lookago Current		_	0.01	1.00	μA	\/ - 20\/				
Output Leakage Current		_	_	2.00	μA	V <sub>OH</sub> = 30V				
Output Low Voltage (Flag)		_	150	250	mV	\\ - 4.5\\				
Output Low Voltage (Flag)		_	_	400	mV	V <sub>IN</sub> = 4.5V, I <sub>OL</sub> = 200μA				
Linner Threshold Voltage		40	60	_	mV	Note 6				
Upper Threshold Voltage		25		_	mV	Note o				
Lower Threshold Voltage		_	75	95	mV	Note 6				
Lower Threshold Voltage		_	_	140	mV	NOIE 0				
Hysteresis		_	15	_	mV	Note 6				

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  - 7:  $V_{REF} \le V_{OUT} \le (V_{IN} 1 \ V)$ , 2.3V  $\le V_{IN} \le 30V$ , 100  $\mu A < I_L \le 100 \ mA$ ,  $T_J \le T_{JMAX}$ .
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Parameters	Sym.	Min.	Тур.	Max.	Units	Conditions			
		_	1.3	_	V	LP2951-02 (±0.5%)			
		_	_	0.7	V	Low			
		2.0	_	_	V	High			
		_	1.3	_	V	LP2951-03 (±1%)			
Input Logic Voltage		_	_	0.7	V	Low			
		2.0	_	_	V	High			
		_	1.3	_	V	LP2951-4.8 (±1%)			
		_	_	0.7	V	Low			
		2.0	_	_	V	High			

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  - 3: Line regulation for the LP2951 is tested at 150°C for  $I_L = 1$  mA. For  $I_L = 100 \mu A$  and  $T_J = 125$ °C, line regulation is guaranteed by design to 0.2%. See Typical Performance Characteristics for line regulation versus temperature and load current.
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Parameters Sym. Min. Typ. Max. Units Condition									
			30	50	μA	- 2 4)/			
Object descent Occurrent		_	_	100	μA	$V_{SHUTDOWN} = 2.4V$			
Shutdown Input Current		_	450	600	μA	- 201/			
			_	700	μA	V <sub>SHUTDOWN</sub> = 30V			
Regulator Output Current in Shutdown		_	3	10	μA	Nete O			
		_	_	20	μA	Note 9			

- **Note 1:** Output or reference voltage temperature coefficient is defined as the worst case voltage change divided by the total temperature range.
  - 2: Regulation is measured at constant junction temperature, using pulse testing with a low duty cycle. Changes in output voltage due to heating effects are covered in the specification for thermal regulation.
  - 3: Line regulation for the LP2951 is tested at 150°C for I<sub>L</sub> = 1 mA. For I<sub>L</sub> = 100 μA and T<sub>J</sub> = 125°C, line regulation is guaranteed by design to 0.2%. See Typical Performance Characteristics for line regulation versus temperature and load current.
  - **4:** Dropout voltage is defined as the input to output differential at which the output voltage drops 100 mV below its nominal value measured at 1V differential. At very low values of programmed output voltage, the minimum input supply voltage of 2V (2.3V over temperature) must be taken into account.
  - 5: Thermal regulation is defined as the change in output voltage at a time T after a change in power dissipation is applied, excluding load or line regulation effects. Specifications are for a 50 mA load pulse at V<sub>IN</sub> = 30V (1.25W pulse) for t = 10 ms.
  - **6:** Comparator thresholds are expressed in terms of a voltage differential at the Feedback terminal below the nominal reference voltage measured at 6V input. To express these thresholds in terms of output voltage change, multiply by the error amplifier gain = V<sub>OUT</sub>/V<sub>REF</sub> =(R1 + R2)/R2. For example, at a programmed output voltage of 5V, the Error output is guaranteed to go low when the output drops by 95 mV x 5V/1.235V = 384 mV. Thresholds remain constant as a percent of V<sub>OUT</sub> as V<sub>OUT</sub> is varied, with the dropout warning occurring at typically 5% below nominal, 7.5% guaranteed.
  - 7:  $V_{REF} \le V_{OUT} \le (V_{IN} 1 \ V)$ , 2.3V  $\le V_{IN} \le 30V$ , 100  $\mu A < I_L \le 100 \ mA$ ,  $T_J \le T_{JMAX}$ .
  - **8:** Output or reference voltage temperature coefficient is defined as the worst case voltage change divided by the total temperature range.
  - 9:  $V_{SHUTDOWN} \ge 2V$ ,  $V_{IN} \le 30V$ ,  $V_{OUT} = 0$ , with Feedback pin tied to 5V Tap.

# LP2951

# **TEMPERATURE SPECIFICATIONS (Note 1)**

Parameters	Sym.	Min.	Тур.	Max.	Units	Conditions				
Temperature Ranges										
Lead Temperature Range	_	_	260	_	°C					
Junction Operating Temperature	TJ	-40	_	+125	°C					
Storage Temperature Range	T <sub>A</sub>	-65	_	+125	°C					
Package Thermal Resistances	•	•			•					
Thermal Resistance, DIP Package	$\theta_{\sf JA}$	_	105	_	°C/W	Soldered directly to a PC board				
	$\theta_{JC}$	_	40	_	°C/W					
Thermal Resistance, SOIC Package	$\theta_{JA}$	_	160	_	°C/W	Typically mounting on a 1" square copper-clad FR4 circuit board				
	$\theta_{\sf JC}$	_	25	_	°C/W					

Note 1: The maximum allowable power dissipation is a function of ambient temperature, the maximum allowable junction temperature and the thermal resistance from junction to air (i.e., T<sub>A</sub>, T<sub>J</sub>, θ<sub>JA</sub>). Exceeding the maximum allowable power dissipation will cause the device operating junction temperature to exceed the maximum +125°C rating. Sustained junction temperatures above +125°C can impact the device reliability.

#### 2.0 TYPICAL PERFORMANCE CURVES

**Note:** The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore outside the warranted range.

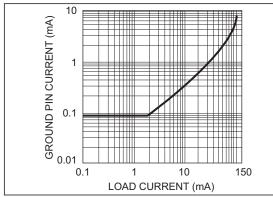


FIGURE 2-1: Quiescent Current.

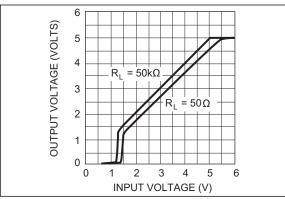


FIGURE 2-2: Dropout Characteristics.

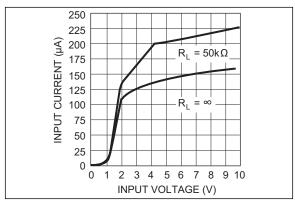


FIGURE 2-3: Input Current.

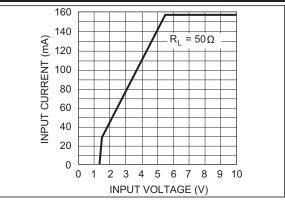
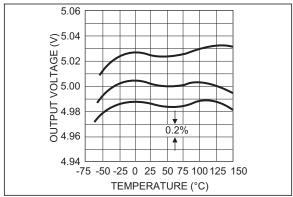


FIGURE 2-4: Input Current.



**FIGURE 2-5:** Output Voltage vs. Temperature of 3 Representative Units.

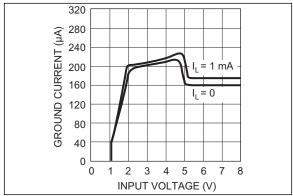


FIGURE 2-6: Ground Current.

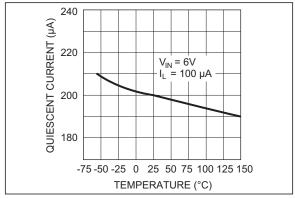
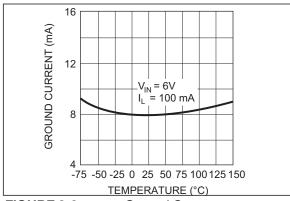


FIGURE 2-7: Quiescent Current.



**FIGURE 2-8:** Ground Current.

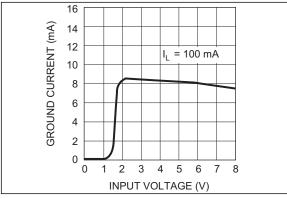
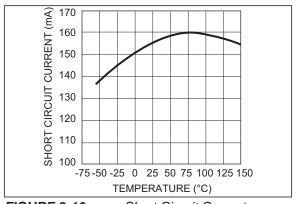
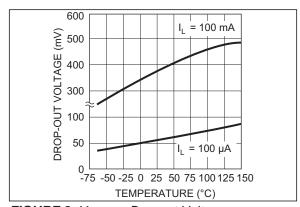


FIGURE 2-9: Ground Current.

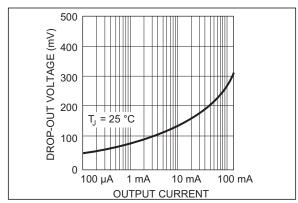


**FIGURE 2-10:** 

Short Circuit Current.

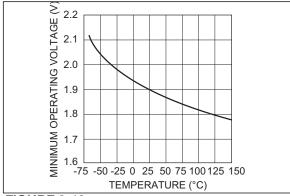


**FIGURE 2-11:** Dropout Voltage.

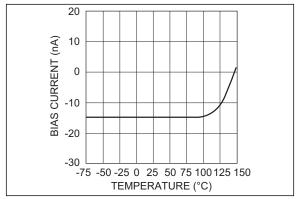


**FIGURE 2-12:** 

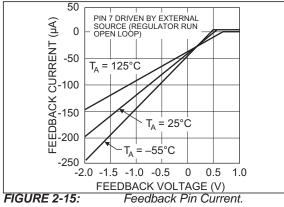
Dropout Voltage.



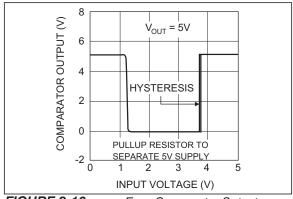
**FIGURE 2-13:** Minimum Operating Voltage.



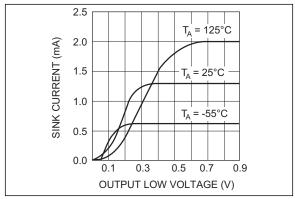
**FIGURE 2-14:** Feedback Bias Current.



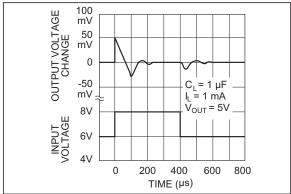
**FIGURE 2-15:** 



**FIGURE 2-16:** Error Comparator Output.



**FIGURE 2-17:** Comparator Sink Current.



**FIGURE 2-18:** Line Transient Response.

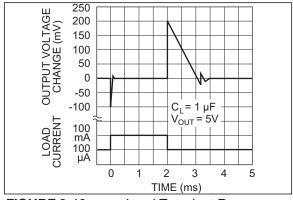


FIGURE 2-19: Load Transient Response.

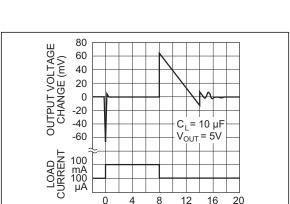


FIGURE 2-20: Load Transient Response.

TIME (ms)

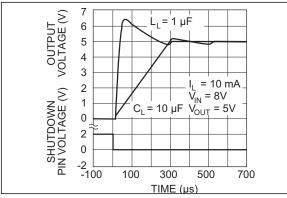


FIGURE 2-21: Enable Transient.

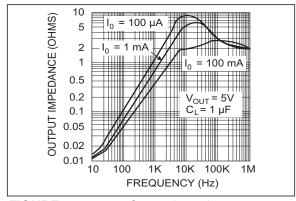


FIGURE 2-22: Output Impedance.

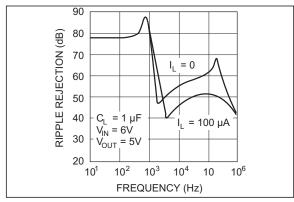


FIGURE 2-23: Ripple Rejection.

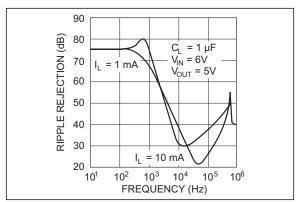


FIGURE 2-24: Ripple Rejection.

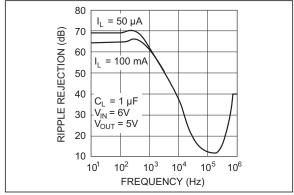


FIGURE 2-25: Ripple Rejection.

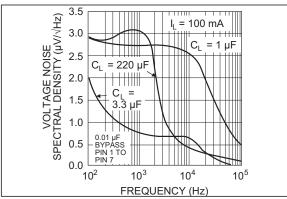


FIGURE 2-26: Output Noise.

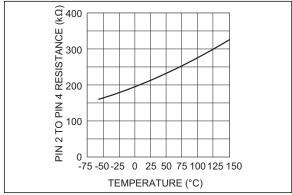


FIGURE 2-27: Divider Resistance.

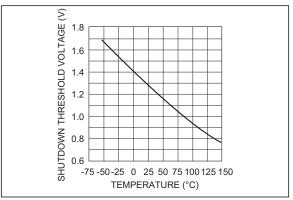


FIGURE 2-28: Shutdown Threshold Voltage.

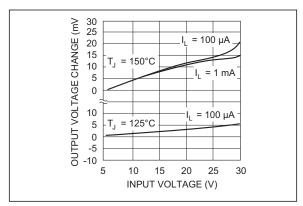


FIGURE 2-29: Line Regulation.

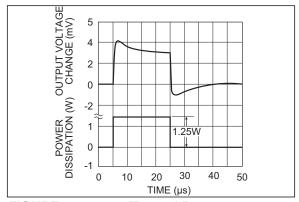


FIGURE 2-30: Thermal Response.

#### 3.0 PIN DESCRIPTIONS

The descriptions of the pins are listed in Table 3-1.

TABLE 3-1: DIP AND SOIC PIN FUNCTION TABLE

LP2951	Symbol	Description			
1	V <sub>OUT</sub>	Regulated voltage output.			
2	SENSE	Output Voltage Sense.			
3	SHDN	Shutdown Input.			
4	GND	Ground Terminal.			
5	ERROR	Error Output.			
6	5V TAP	Internal Resistor Divider for 5V Output.			
7	FB	Voltage Feedback Input.			
8	V <sub>IN</sub>	Unregulated Supply Voltage.			

#### 4.0 APPLICATION INFORMATION

#### 4.1 External Capacitors

A 1.0  $\mu$ F (or greater) capacitor is required between the LP2951 output and ground to prevent oscillations due to instability. Most types of tantalum or aluminum electrolytics will be adequate; film types will work, but are costly and therefore not recommended. Many aluminum electrolytics have electrolytes that freeze at about  $-30^{\circ}$ C, so solid tantalum capacitors are recommended for operation below  $-25^{\circ}$ C. The important parameters of the capacitor are an effective series resistance of about  $5\Omega$  or less and a resonant frequency above 500 kHz. The value of this capacitor may be increased without limit.

At lower values of output current, less output capacitance is required for output stability. The capacitor can be reduced to 0.33  $\mu\text{F}$  for current below 10 mA or 0.1  $\mu\text{F}$  for currents below 1 mA. Using the 8-pin versions at voltages below 5V runs the error amplifier at lower gains so that more output capacitance is needed. For the worst-case situation of a 100 mA load at 1.23V output (Output shorted to Feedback) a 3.3  $\mu\text{F}$  (or greater) capacitor should be used.

When the 5V Tap pin and Feedback pin are connected together for 5V output voltage, the LP2951 will remain stable and in regulation with no load in addition to the internal voltage divider, unlike many other voltage regulators. This is especially important in CMOS RAM keep-alive applications. When setting the output voltage of the LP2951 with external resistors, a minimum load of 1  $\mu\text{A}$  is recommended.

A 0.1  $\mu$ F capacitor should be placed from the LP2951 input to ground if there is more than 10 inches of wire between the input and the AC filter capacitor or if a battery is used as the input.

Stray capacitance to the LP2951 Feedback terminal (pin 7) can cause instability. This may especially be a problem when using high value external resistors to set the output voltage. Adding a 100 pF capacitor between Output and Feedback and increasing the output capacitor to at least 3.3  $\mu\text{F}$  will remedy this.

# 4.2 Error Detection Comparator Output

A logic low output will be produced by the comparator whenever the LP2951 output falls out of regulation by more than approximately 5%. This figure is the comparator's built-in offset of about 60mV divided by the 1.235V reference voltage. (Refer to the block diagram on Page 1). This trip level remains "5% below normal" regardless of the programmed output voltage of the LP2951. For example, the error flag trip level is typically 4.75V for a 5V output or 11.4V for a 12V

output. The out of regulation condition may be due either to low input voltage, current limiting, or thermal limiting.

Figure 4-1 is a timing diagram depicting the ERROR signal and the regulated output voltage as the LP2951 input is ramped up and down. The ERROR signal becomes valid (low) at about 1.3V input. It goes high at about 5V input (the input voltage at which  $V_{OUT}$  = 4.75V). Since the LP2951's dropout voltage is load-dependent (see curve in Typical Performance Curves), the input voltage trip point (about 5V) will vary with the load current. The output voltage trip point (approximately 4.75V) does not vary with load.

The error comparator has an open-collector output which requires an external pull-up resistor. Depending on system requirements, this resistor may be returned to the 5V output or some other supply voltage. In determining a value for this resistor, note that while the output is rated to sink 400  $\mu A$ , this sink current adds to battery drain in a low battery condition. Suggested values range from 100 k $\Omega$  to 1 M $\Omega$ . The resistor is not required if this output is unused.

#### 4.3 Programming the Output Voltage

The LP2951 may be pin-strapped for 5V output voltage using its internal voltage divider, by tying Pin 1 ( $V_{OUT}$ ) and Pin 2 (SENSE) together, as well as tying Pin 7 (Feedback) and Pin 6 (5V TAP) together. Alternatively, it may be programmed for any output voltage between its 1.235V reference and its 30V maximum rating. An external pair of resistors is required, as shown in Figure 4-2.

The complete equation for the output voltage is:

#### **EQUATION 4-1:**

$$V_{OUT} = V_{REF} \times \left\{1 + \frac{R_1}{R_2}\right\} + I_{FB}R_2$$

Where:

V<sub>REF</sub> = the nominal 1.235 reference voltage I<sub>FB</sub> = the feedback pin bias current, nominally 20 nA

The minimum recommended load current of 1  $\mu$ A forces an upper limit of 1.2 M $\Omega$  on the value of R2, if the regulator must work with no load (a condition often found in CMOS in standby), I<sub>FB</sub> will produce a 2% typical error in V<sub>OUT</sub> which may be eliminated at room temperature by trimming R1. For better accuracy, choosing R2 = 100 k $\Omega$  reduces this error to 0.17% while increasing the resistor program current to 12  $\mu$ A. Since the LP2951 typically draws 60  $\mu$ A at no load with Pin 2 open-circuited, this is a small price to pay.

#### 4.4 Reducing Output Noise

In reference applications it may be advantageous to reduce the AC noise present at the output. One method is to reduce the regulator bandwidth by increasing the size of the output capacitor. This method is relatively inefficient, as increasing the capacitor from  $1\mu F$  to  $220~\mu F$  only decreases the noise from  $430~\mu V$  to  $160~\mu V_{rms}$  for a 100~kHz bandwidth at 5V output.

Noise can be reduced fourfold by a bypass capacitor across R1, since it reduces the high frequency gain from 4 to unity. Pick the resulting frequency from Equation 4-2:

#### **EQUATION 4-2:**

$$C_{BYPASS} = \frac{1}{2\pi R_1 \bullet 200 Hz}$$

or about 0.01 µF. When doing this, the output capacitor must be increased to 3.3 µF to maintain stability. These changes reduce the output noise from 430 µV to 100 µV $_{rms}$  for a100 kHz bandwidth at 5V output. With the bypass capacitor added, noise no longer scales with output voltage so that improvements are more dramatic at higher output voltages.

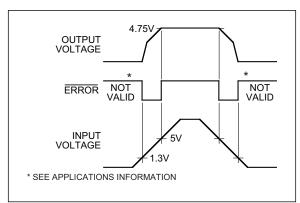


FIGURE 4-1: ERROR Output Timing.

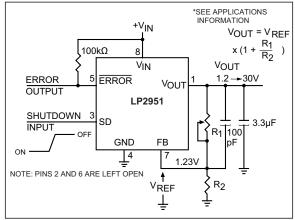


FIGURE 4-2: Adjustable Regulator.

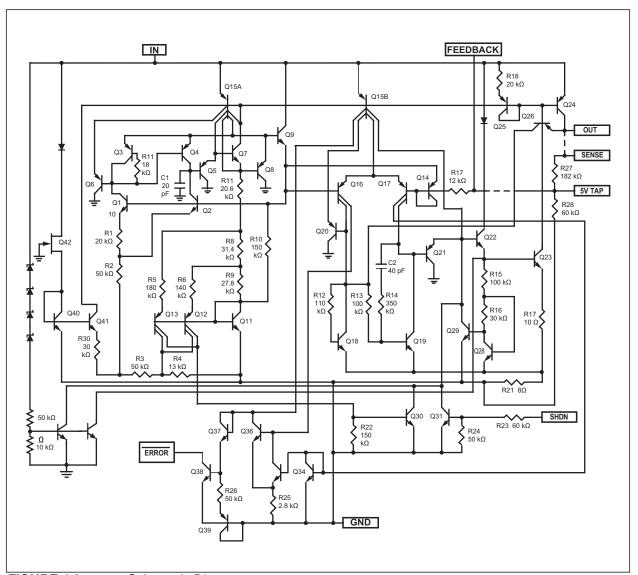
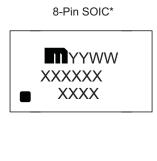
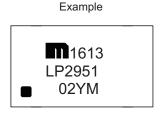


FIGURE 4-3: Schematic Diagram.

#### 5.0 PACKAGING INFORMATION

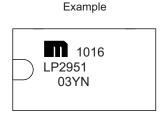
#### 5.1 Package Marking Information





YYWW XXXXXX XXXX

8-Pin Plastic DIP\*

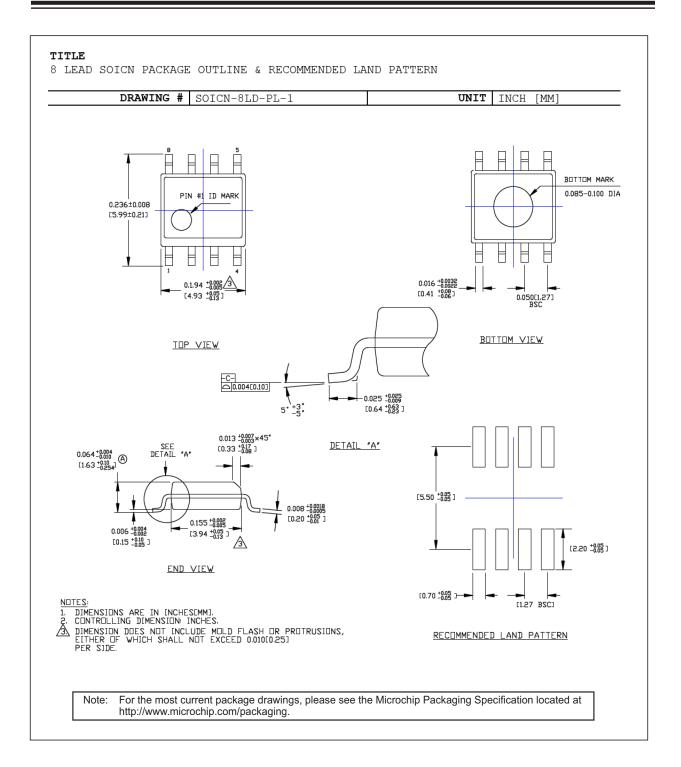


Legend: XX...XProduct code or customer-specific information Υ Year code (last digit of calendar year) YY Year code (last 2 digits of calendar year) WW Week code (week of January 1 is week '01') NNN Alphanumeric traceability code Pb-free JEDEC® designator for Matte Tin (Sn) **e**3 This package is Pb-free. The Pb-free JEDEC designator (@3)) can be found on the outer packaging for this package. •, ▲, ▼ Pin one index is identified by a dot, delta up, or delta down (triangle mark).

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the corporate logo.

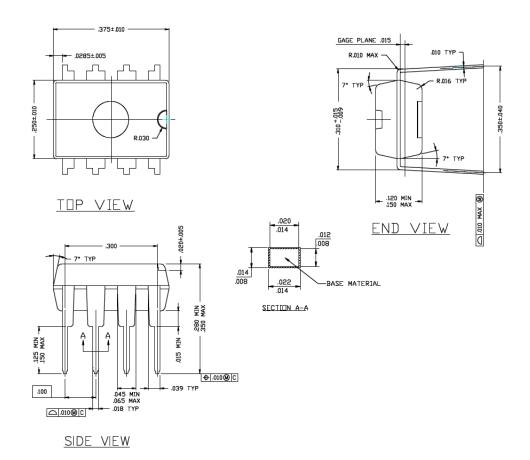
Underbar ( ) and/or Overbar ( ) symbol may not be to scale.



#### TITLE

8 LEAD PDIP PACKAGE OUTLINE & RECOMMENDED LAND PATTERN

DRAWING #	PDIP-8LD-PL-1	UNIT	INCH
Lead Frame	Copper	Lead Finish	Matte Tin



Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging.

#### APPENDIX A: REVISION HISTORY

## Revision A (May 2017)

- Converted Micrel document LP2951 to Microchip data sheet template DS20005736A.
- Minor grammatical text changes throughout.

# LP2951

**NOTES:** 

## PRODUCT IDENTIFICATION SYSTEM

To order or obtain information, e.g., on pricing or delivery, contact your local Microchip representative or sales office.

PART	NO.	- <u>X</u>	X	Exa	amples:	
Device:		turacy Temperature Range	Package  /oltage Regulator	a)	LP2951-02YM:	100 mA Low-Dropout Voltage Regulator, 0.5% Accuracy, —40°C to +85°C (RoHS Compliant) 8LD SOIC
Accuracy: Temperature Range:		= 0.5% = 1.0% = -40°C to +85°C (RoHS)	Compliant)	b)	LP2951-03YM:	100 mA Low-Dropout Voltage Regulator, 1.0% Accuracy, -40°C to +85°C (RoHS Compliant) 8LD SOIC
Packages:		= 8-pin SOIC = 8-pin DIP		c)	LP2951-02YN:	100 mA Low-Dropout Voltage Regulator, 0.5% Accuracy, -40°C to +85°C (RoHS Compliant) 8LD DIP
				Not	catalog part identifier is u is not printed with your Mic	el identifier only appears in the number description. This sed for ordering purposes and on the device package. Check rochip Sales Office for package ith the Tape and Reel option.

# LP2951

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